

TRANSIENT RESPONSES IN MULTIPLE PLATE ABSORPTION COLUMNS

by

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May 20, 1960

Professor Philip Franklin Secretary of the Faculty Massachusetts Institute of Technology Cambridge 39, Massachusetts

Dear Frofessor Franklin:

In accordance with the requirements for zraduation, I herewith submit <sup>a</sup> thesis entitled "Transient Responses in Multiple Plate Absorption Columns"

Sincerely yours,

# Signature redacted

William C. Morris

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## I, SUMMARY

In order to gain some understanding of the unsteadystate response of multiple plate absorption towers, this thesis uses a Reeves Electronic Analogue Computer to duplicate the changes in plate liquid concentration of an absorber whose vapor feed concentration has experienced a step change. Agsumptions used in formulating the tower model Include: theoretical plates, no concentration gradient across the plates, constant inert liquid and vapor flow rates, negligible vapor holdup,. and <sup>a</sup> linear equilibrium relstionship.

It was found that after an input step the liquid concentration on each plate approached a new steady-state value slong a path which could be closely epproxlimated by a time leg (or jump) followed by an exponential rise.

The time lag was different on each plate of <sup>a</sup> tower, being largest (and positive) on the top plate and smallest (negative) on the bottom plate. The leg wes independent of the size of the input disturbance.

The exponential was independent of both the plate position and the size of the input step. It is suggested that the flrst exponent in the analytical solution of Lapidus and Amundson (5) can be used to approximate the predominant exponent  $s_i = \frac{-VK}{h} \left[ \frac{L}{VK} + 1 - 2 V_{VK} - \cos \frac{\pi}{N+1} \right]$ 

 $-1-$ 

#### II. INTRODUCTION

Treditionally, chemical engineers have focused thelr attention on the steady-state characteristics of chemical processes. As the sophistication of steady-state techniques hag grown, however, Increasing attentlon has been given to the time dependent aspects of process response.

Basically, there are two general types of unsteady-state phenomena: the response of a system to a eyclliely changing input, termed frequency response, and the response of <sup>a</sup> system to a non-repetitive input disturbance, termed transient response. Both of these general response characteristics are of importance in chemical process equipment, but of the two, transient response is both easier to grasp conceptually, and more valuable from an elementary standpoint.

<sup>A</sup> particular case of interest from the transient point of view is the multiple plate fractionating column. In such a column a binary feed 1s introduced somewhere between the top and bottom plate, the volatile component taken out the top, and the non-volatile component off the bottom plate. Normally heat 1s added to the bottom plate, or still, and some fraction of the top vapor is condensed and returned es liguid reflux to the top plate. Obviously, the reflux ratio, and the rate at which heat 1s added will considerably affect the purity of the product streams. Steady-state methods enable one to calculate the composition of the product streams, given fixed values of feed composition,

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reflux ratio, heat supplied, equilibfium data, and other tower variables, If such <sup>a</sup> tower is once characterized, and <sup>a</sup> discontinuous change (step change) is made in <sup>a</sup> parameter such as feed composition, then the same steadystate technique will furnish the new equilibrium values of product concentration, It will furnish nothing, however, about the time necessary for the output to equilibrate, nor will it define the path taken by the output in reaching its new equilibrium values,

The usefulness of being able to define the transsient response for control of <sup>a</sup> column such as this is pointed out by Gilliland (9). It is often desired to control the reflux ratio of such <sup>a</sup> tower by means of <sup>a</sup> servomechanism coupled to a sensing device which monitors the compo= sition of the material in the tower. The object is to control the reflux ratio in such a manner as to isolate the top product from disturbances in operation parameters such as feed composition. Unfortunately, however, sensing devices able to withstand the rigors of plant operation are not capable of detecting the minute swings in purity which occur at the ton of such <sup>a</sup> tower. For this reason the device 1s ususlly placed further down in the tower where concentration swings are proportionally larger. Thus a lag occurs between the time a disturbance resches the sensing device, and the time it reaches the top of the tower. In this case, the sensing device can operate on the reflux only through some functional lag operator which depends on the column's input-

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output transient response as well as its input-sensing device transient response.

Work in this field has been going on for some time, At M.I.T., Smith and Polk  $(12)$ , Jordan  $(4)$ , O'Donnel  $(8)$ , and Davis (2) have all examined the transient resvonse of <sup>a</sup> fractionating column with this control oroblem in mind, In every cese they developed mathematical models of a column and solved for the transient solution on <sup>a</sup> digital computor.. Davis was even able to duplicate the reflux control problem, and hold product composition constant while supplying a step feed change. Jackson and Pigford (3) cherscterized the transient responses of columns with both very large and very small reboilers. Their work was also done on <sup>a</sup> digital computor. Rose, Johnson, and Williems (10,11) examined both column control and general column transients on digital and analogue computers. Almost all of this previous work assumed non-linear equilibrium relationships in construstion of the models, While specific solutions to the cases studied were obtained, attempts to correlate the data for general applicability were not so successful. In sddition, the effect of changes in the opersting end design parameters were complicsted and difficult to understand.

In an attempt to clarify the existing results, this project reverts to the examination of a simpler- but relatedcase, the case of <sup>g</sup> linear equilibrium relationship and half of <sup>a</sup> fractionating tower. These conditions

correspond to a multiple plate absorber.

Marshall end Pigford (6) first derived the solution to <sup>a</sup> set of lineer sbsorber equations, Lapidus and Amundsen (5) extended this treatment to cases with arbitrary and conditions. Their result is ststed in the form of <sup>a</sup> set of sums and integrals whose solution is involved for many cesses, Acrivos and Amundson (l) examined gbsorbers with non-linear equilibrium reletionshivs on an anslogue computor. No correlations were presented in any of this work to establish the effect of parameters on the transients,

This project also examines the transient response of an absorption tower by means of an analogue computor. Linear equilibrium relstionships are used snd an attempt mede to recognise the effect of the various design and operating varisbles on the transient times,

<sup>A</sup> multiple plate absorber consists of <sup>a</sup> tower filled with plates which contain either smsll holes or bubble caps to allow the upward passage of vapor. The top side of the plates is covered with liquid absorbant which flows across the plate snd down an overXlow to the plate below, This 1iquid ebsorbant 1s introduced at the top of the tower and allowed to mix with the rising vapor until it emerges from the bottom of the unit enriched in acquired solute. The vapor feed with a large solute concentration is introduced at the bottom of the tower, and passed upward through the plates until 1t emerges from the top stripped to some fraction of its initial solute concentration.

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Examining an arbitary absorber tower such as 1s shown in Figure 1, and assuming theoretical plates, no concentration gradient on the plates, and no mass transfer other than solute between the streams, it is possible to relate the <sup>c</sup> change in any plate concentration to the material flows to that plate by the expression:

$$
V_{\overline{y}_{n+1}} + Lx_{n-1} - Vy_n - Lx_n = h \frac{\partial x_n}{\partial \theta} - H \frac{\partial y_n}{\partial \theta}
$$
 (1)  
here: V = The solute free vapor rate, moles/time  
L = The solute free liquid rate, moles/time  
x = moles solute/mole solute free liquid  
y = moles solute/mole solute free vapor  
h = liquid holdup per plate  
H = vapor holding per plate

Assuming a negligible vapor holdup (a reasonable condition in the case of an sbsorber), and <sup>a</sup> linear equilibrium expression, Equation  $(1)$  reduces to:

$$
\frac{VK}{h}X_{n+1} + \frac{L}{h}X_{n-1} - \left(\frac{L+VK}{h}\right)X_n = \frac{\partial X_n}{\partial \theta}
$$
 (2)

where K is defined by the equilibrium relationship:<br>  $y_n = Kx_n$ 

$$
y_n = Kx_n \tag{3}
$$

Equation  $(2)$  is the general form of N equations corresponding to the N plates of any given absorber.

In this work. three sets of such equations are solved for stepped feed composition on the Reeves Analogue Computer located at M.I.T. The effects of the parameters V, L, and Ay, and the design parameter <sup>N</sup> on <sup>a</sup> transient resulting from stepped input concentration are analysed.



#### III. PROCEDURE

## A. Conditions Studied

In orded to evaluate the effects of the parameters  $L$ ,  $V$ ,  $\Delta y$  and N on absorber response three absorber models were studied.

The first model, used in Series 2, had five plates, <sup>a</sup> vapor rate of <sup>25</sup> moles/hour, <sup>a</sup> liquid rate of <sup>5</sup> moles/hour, and <sup>a</sup> liquid plate holdup of <sup>1</sup> mole/plzte. In this, as well as the subsequent series, a negligible amount of vapor holdup. and an equilibrium constant of 0.20 was assumed, The second model, used in Series 3, had six plates. Otherwise the assumed parameters were ldentical with the model in Series 2. The third model,. Series 4, had six plates, and <sup>a</sup> liquid rate of <sup>6</sup> moles/hour. In all other respects 1t was 1dentical with the absorber in Series 2. Series 1, which used the same model as Series 2. gave results which were discarded from the analysis because of unsophisticated computer programing techniques. Table I is <sup>a</sup> summary of parameters in the three series.

## B. The Reeves Analogue Computer

The Reeves Analogue Computer (REAC) duplicates variables in an ecuation with voltages and time. A sub-circuit is set up for each plate equation, and linked with appropriate positions in the preceeding and following plate sub-circuits. Initial conditions are fed to the circuit in the form of voltages,.andthenetwork 1s allowed to equilibrate through the flow of currents. <sup>A</sup> solution to the transient response 1s obtained by measuring the change in voltage with

time at an appropriate place in the circuit,

The primary components of the computer are high gain d.c. smplifiers arranged to perform the functions of integration, summation, and inversion. The Reves Electronic Analogue Computer (REAC) located at M.I.T. contains twenty of these d.c. amplifiers, Seven are used as summers, seven as integrators, and six as inverters, As long as an sbsorber with linear characteristies is assumed, only summers, integrators and potentiometers are needed to duplicate the »TOblem.

#### TABLE I

Summary of Values Used in Absorber Models



The circuit components in the REAC are all individually wired to a patch-bay located on the front of the computer. This bay is designed to accept <sup>a</sup> pre-patch-board on which <sup>a</sup> combination of computer components can be connected by plug-in leads. The pre-patch-board can be removed from the machine without disturbing the problem patched onto it. Thus, one problem need not be destroyed before another can be run.

 $-8-$ 

51x of the integrators, whose inputs and outputs are wired to the patch bay, have Intermal circults which permit initial conditions in the form of constant voltages to be added to thelr outputs, These initlal conditions are set with six potentiometers (or simply pots) on the computer front. Associated with each pot is <sup>a</sup> switch which permits 1ts output to be of elther algebraic sign.

The high gain emplifiers used in both the integrators and summers are designed to operate with their outputs in <sup>a</sup> range of  $±100$  volts. Driving the element outside this range produces nonlinearity, and neon overload lights are provided to indicate this. All of the amplifiers in the computer sive rise to <sup>a</sup> sign change between thelr outputs and their inputs.

Any two voltages in the computer can be monitored with the externally mounted Output Table. This table consists of a cylindrical drum on which graph paper is mounted, and a traversing arm (with a pen holder) which moves across the outside of the drum in <sup>2</sup> line persllel with its axis. Both the rotation of the drum, and the displacement of the arm are controlled with servomechenisms that derive their positioning signals from any desired point in the computer Circuit. For <sup>a</sup> more detziled discussion of computer com= ponents and technlgues the resder is referred to the REAC Theory and Operation Manual  $(7)$ .

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## III. PROCEDURE

# 0. Programing and Operating the Computer

# Scaling the Problem

In order to convert the absorber model from the moles/mole -hour domain to the voltage-second domain of the computer, two changes of variables must be made. To relate <sup>x</sup> with voltage (e) the factor <sup>C</sup> is chosen so that

$$
e_n = Cx_n \qquad (4)
$$

To speed up the problem solution, and to stay within the dynamic range of the computer a factor is used in the time scale.

$$
t = a\theta
$$
 (5)  
where  $t =$  computer time (in seconds)  
 $\theta =$  physical time (in hours)  
 $a =$ scale factor

Substituting these two variable changes (4 and 5) into (2) ve arrive at <sup>a</sup> computer equation,

$$
\frac{VK}{ha}e_{n+1} + \frac{L}{ha}e_{n+1} - \left(\frac{VK+L}{ha}\right)e_n = \frac{\partial e_n}{\partial t}
$$
 (6)

vhich corresponds to the physical problem.

Since <sup>C</sup> does not appear in Ecuation (6), the scaler factor relating x and <sup>e</sup> does not affect solution time, and may be choosen with regard only for full utilization of the voltage range in the machine. As most electrical devices the computer is most accurate when opperated near its maximum voltage limit. The time factor a, however, directly influences the solution time by operating on the constants applied to each integrator output. In all the runs of this thesls the following values were used for <sup>a</sup> and C:  $= 20$   $C = 300$ 

 $a = 20$ 

## C. Programing end Operating the Computer

#### Patching In the Froblem

Figure <sup>2</sup> shows the complete block diagram of Series <sup>3</sup> on the computer. Because the solute concentration in the liquid feed 1s zero, the only voltage input to the network corresponds to the feed vapor solute concentration entering plate (integrator) number 5. The initial condition circuits simply boost each plate to near steady-state value and shorten the time necessary for equilibration of the computer before the step is introduced. Because of the sign change in the amplifiers, consecutlve integrators alternate in the algebralc sign of their outputs.

<sup>A</sup> step change in vapor feed concentration is accomplished by switching rapidly from Pot 13, which provides the initial  $y_F$ , to Pot 16 which is set at the new value of  $y_F$ . A double-throw function switch is provided on the computer to facilitate such a change. The inputs of Pot 13 and Pot 16 are connected to a 100 volt supply on the patch board. The pots operating on each integrator's output correspond to the constants  $\frac{VK}{h a}$ ,  $\frac{L}{h a}$ , and  $\frac{VK + L}{h a}$  in Equation (6). In Series 3 only two pots per plate were needed because  $\frac{VK}{h\overline{a}}$ <br>was numerically equal to  $\frac{L}{h\overline{a}}$ .

Each integrator's output has <sup>a</sup> potential corresponding, to the licuid solute concentration on thet plate. These outputs are fed consecutively to Summer <sup>1</sup> which 1s used to amplify the voltage by <sup>a</sup> factor of ten before sending 1t to the arm of the Output Table. The high gain of Summer 1 which is necessary for accuracy in the results reculres the



output to be greater than £100 volts if the plate voltage is larger than <sup>±</sup>10 volts. In order to circumvent this problem (i.e. to amplify increments in plate voltage and not its D.C. value) <sup>a</sup> compensating voltege of opposite polarity 1s added to the input of Summer 1. Thus, the displacements registered from zero on the Output Table have no absolute significance and are valuable for indicating changes only.

To provide <sup>a</sup> time scele for the drum rotation (x-axis) of the Output Table <sup>a</sup> one volt signal 1s supplied to Integrator 6 by Pot 15. This integrator thus generates <sup>a</sup> remp of one volt per second which rotates the Output Table's drum at <sup>a</sup> constant velocity. Because the Output Table reads from -100 to +100 volts, the initial condition pot on Integrator <sup>6</sup> 1s used to position the drum at its -100 volt extreme at the beginning of a run.

## Making a Run

Each of the three series of runs made for this study used <sup>a</sup> different absorber model. Individual runs in each series differed only in the magnitude of the step change in their vapor feed concentration.

The first step in making a run was determining the physical parameters of the absorber model. Having these, the machine parameters and constants could be determined from Equations (4) and (5). Once the machine constants were known, the problem was patched into the computer (see Figure 2), and he pots and initial conditions set at their proper values.

 $-12-$ 

Following this, graph paper and pen were placed on the Output Table which had its y-axis servo connected through a summer to one of the integrator outputs and its x-axis servo to a ramp producing integrator. The Output Table axies were then positioned with Pot <sup>19</sup> and the initial condition pot on Integrator <sup>6</sup> in order to prevent the servos from receiving more than <sup>±100</sup> volts during the run.

After placing the function switch in contact with Pot 13, the operating switch, which closes the integrator input connections wes closed and the run started. The Output Table pen plotted the selected plate voltage as <sup>a</sup> function of time. Normally initial conditions were not set with complete accuracy. and several seconds had to be allowed for the machine to equillbrate before introducing the step in yp. Once this egullibrating had occurred, the function switch was used to connect Pot <sup>16</sup> with the circuit. The effect of the step in the monitored integrator (as well as the preceeding equllibration changes) was graphed on the Output Table. In order to complete <sup>a</sup> run with data from each integrator, this procedure hed to be repeated once with the Output Table connected to each plate. No significant error was introduced by this technique (see DISCUSSION OF RESULTS, C. Accuracy of Results).

The five or six curves which corresponded to each plate in the tower were drawn by the Output Table on <sup>a</sup> single sheet of graph paper. Figure <sup>3</sup> is an example of one such data sheet. It should be pointed out that the alternating directions of change following the step are the result of the polarity reversal in each consecutive intezrator output.

 $-13-$ 



#### IV. RESULTS

The data taken from the computer on graphs such as Flgure <sup>3</sup> was reduced to numerical form by measuring the amount of change in  $x_n$  after each three squares on the time axis (about <sup>10</sup> seconds on the computer or 0.5 hours in the tower).

The curves for each plate seemed to rise or fall exponentially from their initial to their final equilibrium values. In order to examine any exponential tendency in the response, the curves were replotted on semi-logarithmic graph paper.

Because <sup>a</sup> rising exponential cannot be devicted as <sup>a</sup> straight line on semi-log paper, the values measured fron the raw data were related to the final equilibrium value rather than the initial equilibrium value, i.e.  $x_n(\infty) - x_n(\theta)$  was measured rather than  $x_n(\theta)-x_n(0)$ . In addition, these values were normalized to facilitate interplate comparison. The final result was taken to be the quantity  $E_p$ , where

$$
E_D = \frac{x_n(\infty) - x_n(\theta)}{x_n(\infty) - x_n(0)}\tag{7}
$$

The curves of log  $E_p$  -vs-  $\theta$  are reproduced in Figures <sup>4</sup> to 20. From these figures it can be seen that each plate in a tower may be approximated by a time lag or jump, where the exponential goes to one at some time other than zero, followed by a single exponential change. It appears from Flgures <sup>4</sup> to <sup>20</sup> that all plates in any given tower have the same time constant for their predominant exponential (i.e.

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 $\Big)$ 





¢





























the slopes of the straight lines on the log  $E_p$  plots are identical. Each plate, however, has <sup>a</sup> different time lag (D) in this simplified approximation.

For every plate the time constant  $(T)$  which is the time required for the exponential to fall to  $\frac{1}{e}$  (or 0.368) of its initial value, and the lag or jump (D) were measured from Figures <sup>4</sup> to 20. Figures 21, 22, and <sup>23</sup> show the distribution of  $\mathcal T$  for each plate in each run of the three series. On any plate of the absorber, the time constants are randomly distributed for different step changes. Furthermore the averages of  $\mathcal T$  for all plates in a series appear to be nearly identical and randomly distributed, l.e. the exponents are the same throughout the absorber regardless of the size or direction of changes in the vapor input concentration.

Figures 24, 25, and 26 show the lag times of each plate for every run in the three series. From these graphs it appears that the lag time of any plate although unicue is not changed by differences in the feed change.

Because of the distributions, the average of all the time constants in each series, and the average of the lags for each plate in all runs of ezch series was taken as <sup>a</sup> best estimate of the true values of these characteristics.











 $r = \mathbb{E} \times \mathbb{E}$ 



Table II 1s <sup>a</sup> tabular summary of the average experimental lags and time constants.

# TABLE II

Summary of Average Values for  $\gamma$  and  $D$ 



Subscripts used to denote plate number

 $\mathcal{L}_{\text{max}}$  and  $\mathcal{L}_{\text{max}}$  . In the  $\mathcal{L}_{\text{max}}$ 

 $\mathcal{L}$ 

#### V. DISCUSSION OF RESULTS

A.. Factors Affecting The Transient Response

As was explained above the time dependent response of each plate in these absorbers may be approximated with a time lag of either sign followed by an exponential decay. It was also pointed out that  $\Upsilon$  and D were independent of the size of the step in vapor feed concentration (see Figures 21 to 26), and that  $\gamma$  was the same for any plate on an absorber. Referring to Table II it is possible to make tentative conclusions about the effects of L, V,  $\Delta y_{\rm m}$ , and N on  $\tilde{J}$  or  $D$ . Table III lists these qualitative effects.

#### TABLE III

## Effect of Parameters On  $\widetilde{J}$  and D



It should be pointed out that the value of  $\frac{L}{V}$  in Series 4 was only slightly different from that used in Series 2 and 3, so 1t 1s difficult to appraise the effect of this change on 7 2nd D.

# V. DISCUSSION OF RESULTS

B. Comparison Of Results With The Analytical Solution

The solution to a linear absorber's equations (2) has been developed by Lapidus and Amundson (5). Their solution is given as:

$$
X_{n}(\theta) = \frac{-2}{N+1} \sum_{k=1}^{N} (-1)^{k} \sin \frac{\pi k (N-n+1)}{N+1} \sum_{j=1}^{n} A_{j}(\sqrt{c})^{n-j} \sin \frac{\pi k j}{N+1} e^{S_{k}\theta}
$$
  

$$
- \frac{2}{N+1} \sum_{k=1}^{N} (-1)^{k} \sin \frac{\pi k n}{N+1} \sum_{j=n+1}^{N} A_{j}(\sqrt{c})^{n-j} \sin \frac{\pi k (N-j+1)}{N+1} e^{S_{k}\theta}
$$
  

$$
- \frac{2(\sqrt{c})^{n-1}}{d(N+1)} \int_{0}^{0} \left[ \frac{3(\theta-b)}{K} \right] \left[ \sum_{k=1}^{N} (-1)^{k} \sin \frac{\pi nk}{N+1} \sin \frac{\pi k}{N+1} e^{S_{k}b} \right] d b
$$

$$
-\frac{z(rc)^{n+1}}{d(N+1)}\int_{0}^{r}f(\theta-t)\left[\sum_{k=1}^{N}(1)^{k}sin\frac{\pi k}{N+1}sin\frac{\pi k(N-n+1)}{N+1}e^{S_{k}b}\right]db\qquad(8)
$$

where:

 $\overline{f}$ 

$$
S_k = \frac{-1}{d} \left[ C + 1 - 2\sqrt{c} \cos \frac{\pi k}{N + 1} \right] \qquad k = 1, 2, \dots, N
$$
\n
$$
C = \frac{L}{\sqrt{K}}
$$
\n
$$
d = \frac{h}{\sqrt{K}}
$$
\n
$$
Y_n = Kx_n
$$
\n
$$
A_j = x_j(o)
$$
\n
$$
f(\theta) = x_o(\theta)
$$
\n
$$
g(\theta) = y_r(\theta)
$$
\n(9)

When  $f(\theta)=0$  and  $g(\theta)=$  Constant, as is the case in this thesis, Equation 8 reduces to:

$$
X_{n}(\theta) = \frac{-2}{N+1} \sum_{k=1}^{N} (-1)^{k} \sin \frac{\pi k (N-n+1)}{N+1} \sum_{j=1}^{n} A_{j} (r c)^{n-j} \sin \frac{\pi k j}{N+1} e^{-s k \theta}
$$
  
\n
$$
+ \frac{2}{N+1} \sum_{k=1}^{N} (-1)^{k} \sin \frac{\pi k n}{N+1} \sum_{j=n+1}^{N} A_{j} (r c)^{n-j} \sin \frac{\pi k (N-j+1)}{N+1} e^{-s k \theta}
$$
  
\n
$$
- \frac{2 (r c)^{n-N}}{d (N+1)} \left( \frac{Y_{r}(\theta > 0)}{K} \right) \sum_{i=1}^{N} (-1)^{k} \sin \frac{\pi n k}{N+1} \sin \frac{\pi k}{N+1} \left( \frac{e^{s k \theta} - 1}{s_{k}} \right)
$$
(10)

When the series are expanded, this expression is a sum of N exponentials whose time constants are determined by  $s_k$ ;

$$
\Upsilon = \frac{-1}{s_k} \tag{11}
$$

The experimental results indicated that one exponential in the solution was predominant. In order to determine which s<sub>k</sub> corresponded to the main exponential, the experimental time constants were compared with the set of  $s_k$ 's for each Table IV lists the results of this comparison. run.

From Table IV it can be seen that the experimental time constants correspond most closely to the theoretical value of  $\int$  for k<sup>-1</sup>. Unfortunately, the diversity in conditions studied was not great enough to provide a more exact evaluation of this correspondence.

It is valuable, however, to examine the nature of Equation 10 as though k=1 were the only term in the expansions. This assumption is reasonably valid in that the small exponents require all other terms to approach zero rapidly.

## TABLE IV

Comparison of Experimental<br>Time Constant With Values of  $s_k$ 



If the only allowable value of k in the exponents is one, Equation 10 becomes:

$$
X_{n}(\theta) = \left[\frac{2}{N+1} \sin \frac{\pi (N-n+1)}{N+1} \sum_{j=1}^{n} A_{j}(\sqrt{c})^{n-j} \sin \frac{\pi j}{N+1} \right] e^{5_{1} \theta}
$$
  
+  $\frac{2}{N+1} \left[ \sin \frac{\pi n}{N+1} \sum_{j=1+n}^{N} A_{j}(\sqrt{c})^{n-j} \sin \frac{\pi (N-j+1)}{N+1} \right] e^{5_{1} \theta}$   
-  $\frac{2(\sqrt{c})^{n-N}}{d(N+1)} \left( \frac{Y_{F}(\theta > 0)}{K} \right) \left[ 51N \frac{\pi n}{N+1} \sin \frac{\pi}{N+1} \right] \frac{e^{5_{1} \theta}}{-5_{1}}$   
+  $\frac{2(\sqrt{c})^{n-N}}{d(N+1)} \left( \frac{Y_{F}(\theta > 0)}{K} \right) \sum_{j=1}^{N} (-1)^{k} S(N \frac{\pi nk}{N+1} \sin \frac{\pi k}{N+1} \left( \frac{1}{+5_{K}} \right)$  (11)

Where all of the k's must be evaluated in the non-exponential integration constant. When  $0=0$  this expression becomes:  $X_n(o) = \left[\frac{2}{N+1} \sin \frac{\pi(N-n+1)}{N+1} \sum_{i=1}^n A_i(\sqrt{c})^{n-j} \sin \frac{\pi j}{N+1}\right] + \left[\frac{2}{N+1} \sin \frac{\pi n}{N+1} \sum_{i=n+1}^N A_i(\sqrt{c})^{n-j} \sin \frac{\pi(N-j+1)}{N+1}\right]$  $-\frac{2(\sqrt{c})^{n-N}}{d(N+l)}\left(\frac{\gamma_F(B>0)}{K}\right)\left[\sin \frac{\pi n}{N+l}\sin \frac{\pi}{N+l}\right]+\frac{2(\sqrt{c})^{n-N}}{d(N+l)}\left[\frac{\gamma_F(B>0)}{K}\right]\left[\frac{\gamma_H(k)}{K}\right]\sin \frac{\pi n k}{N+l}\sin \frac{\pi k}{N+l}\left(\frac{1}{5K}\right)$  $(12)$ 

As 
$$
\theta
$$
 approaches infinity the value of  $x_n$  from Equation 11  
becomes:  

$$
X_n(\infty) = \frac{\partial (\sqrt{c})^{n-N}}{d (N+1)} \left[ \frac{Y_n(\theta > 0)}{K} \right] \left[ \left( -1 \right)^K \sin \frac{\pi nk}{N+1} \sin \frac{\pi k}{N+1} \left( \frac{1}{S_K} \right) \right]
$$
(13)

The path taken by  $x_n$  between these two extremes is determined by the exponential  $e^{S_1\Theta}$ . Thus it appears that the predominate exponential in the step transient has an exponent which corresponds to Equation 9 for  $k=1$ ,

$$
S_{1} = \frac{-VK}{h} \left[ \frac{L}{VK} + 1 - 2V\frac{L}{VK} \cos \frac{\pi}{N+1} \right]
$$
 (14)

Since Equation 11 (and Ecuation 10) reduce to Eguation 13 as  $\theta$  goes to infinity, it would be expected that it gives the final equilibrium value of  $x_n$ . Evaluations of Equation <sup>13</sup> for selected plates in Series <sup>3</sup> are compared with final steady-state values in Table V.

## TABLE V

Comparison Between  $x_n(\omega)$  From Steady-state Analysis and Equation 13

# (Series 3. Run 2)



It appears that the initial values of  $x_n$  given by Equation <sup>12</sup> are not the initial steady-state values. Table VI compares the values of  $x_n(0)$  given by steadystate analysis end Equation 12.

#### TABLE VI

Comparison Between  $x_n(0)$  From Steady-State Analysis and Equation 11

# (Series 3, Run 2)



The differences presented in Table VI may be explained as part of the lag producing mechanism. The value of Equation <sup>12</sup> for the first plate is smaller than the steadystate value. It can be assumed thet the other exponentials counteract this main term (summing to zero) until enough time has elapsed to make it equal to the initial steadystate value. At this point it becomes the predominate term and provides <sup>a</sup> single exponential path to the finsl value.

#### V, DISCUSSION OF RESULTS

#### C.. Accuracy of Results

The accuracy obtainable in the final values of  $\tau$  and <sup>D</sup> for each plate depends primarily on three factors, static accuracy of the computer, dynamic accuracy of the computer, and graphical accuracy in measuring the computer data.

Statle accuracy in the computer can be divided into two significant catigories, individual component accuracy and overall power supply accuracy. As in most electronic equipment precision of components in the REAC cannot be expected to be much less than about 1%, but with carefull compensation of the pots and integrators by <sup>a</sup> vacuum tube volt meter and with computer operation over its full voltage renge, accuracy can be maximized. In addition, loading effects in the pots can be compensated from <sup>a</sup> correction chart (7) or by analyzing their load circuit. The effect of component static accuracy is to change the equilibrium (and transient) voltage on <sup>a</sup> particular integrator output. Since the equilibrium value of the integrator voltage is not read out of the computer, it does not affect the results, and since each plate transient voltage is normalized (from zero to one), its absolute value tends to wash out of  $E_n$ . The deviation of any plate from its actual value is less than about five percent. Inaccuracy in the power supply voltage was large, ranging over long periods about  $\pm 8$  volts from its nominal <sup>100</sup> volt value. This change, however, displaces all the output voltages equally, and since the static values

do not appear in the results, except insofar as they influence dynamic response, an overall static error washes out completely. The time axis was accurately set with <sup>a</sup> volt meter,

Dynamic limitations in the computing elements were negligible In thls thesis. The only elements with narrow bandwidths were integrators, and the runs were made with ample provision for limitations in this type of element.

The most critical error in deriving the results was the accuracy with which the Output Table grephs (see Figure 4) could be read. Two factors affected this accuracy: the preclsion obtained in alligning the graph paper on the drum, and accuracy with which the transient displacement could be read from the graph. The allignment of the graph was accomplished with an accuracy of  $\texttt{\texttt{10.3}}$  small squares per three hours (values from the graph were tabulated in small square units). Since  $x_n(\infty) - x_n(0)$ , expressed in small squares, varied between 10 for plate one and about <sup>150</sup> for plates five end six, this cumulative error ranged from  $\pm 3\%$  to  $\pm 0.3\%$ . The error was largest for plates one snd two.

Displacement from the final equilibrium value could be measured with dividers to an accuracy of about  $10.1$ small squares. The percent error caused at any point by this deviation depended on both the value of  $x_n(\infty) - x_n(0)$ and the degree to which the transient had diminished this

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value, Plates one and two show <sup>a</sup> high percentage error because their initial values were low. Initially, the errors in these plates were on the order of  $\pm 1\%$  (since no run with an  $x_1(\infty) - x_1(0)$  of less than 10 small squares was included in the analysis), but as these first plates approached thelr new equilibrium, the error increased exponentially to about  $\pm 10\%$  (no value smaller than one small square was considered in the results). Thus, accuracy fell off sharply with time. Since the first pletes had low equilibrium changes, and since the last rose rapidly because of its time Jump, the middle plates were the least affected by this limitation. It should be pointed out that the graph displacements were amplified by a factor of ten in order to decrease thls inaccuracy.

In summary, the most important limitatlons on accuracy were caused by imperfect graph allignment and lack of accuracy in measuring small increments. In the range of time where the lines determining  $\mathcal T$  and  $D$  were measured, these limitations caused an error of up to about £10% in the top two plates, about  $\pm 4\%$  in the middle plates, and about  $\pm 6\%$  in the bottom plates. For large increments in  $y_F$  these<br>errors were reduced appreciably.<br>Error in  $\tau$  and D caused by these effects corresponded errors were reduced appreciably.

to the same values, but the averages which were derived as best estimates placed limits of  $\pm 2\%$  on  $\mathcal T$  and about  $\pm 4\%$  on D.

#### VI. CONCLUSIONS

From the results of this thesis the following concluslons may be made about the response of an absorber tower to a step change in the vapor feed concentration, when the tower 1s chacterized by perfect plates, negligible vapor holdup, constant inert liquid and vapor streams, and <sup>a</sup> linear equilibrium relationship.

- l. The resultant chenge in plate concentration may be closely approximated by a time lag followed by en exponential transient.
- The exponent characterizing the exponential is the  $2.$ same in every plate, and independent of the size of the feed disturbance.
- $\overline{3}$ . The transient's exponent may be closely approximated by  $s_i = -\frac{VK}{b} \left[ \frac{L}{VK} + 1 - 2\sqrt{\frac{L}{UK}} \cos \frac{\pi}{N+1} \right]$
- 4. The time lag preceeding the exponential is different on every plate, being largest and positive on the top plate, and smallest (negative) on the bottom plate
- 5. The time lag is independent of the size of the feed disturbance.

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#### VII. RECOMMENDATIONS

As an ald to continuing and expanding the results of this thesis the following recommendations may be made.

- More data on different absorber modles should be  $1.$ run to check the proposed correlation for the time constant(Equation 14). It should be necessary to run only one or two steps on each model if the steps are large enough to produce zood accuracy.
- $2.$ Some investigation should be made into the effect of changing the vapor-liquid ratio in a tower such as the one used in this thesis. Such a chenge (in the liquid rate for example) is the most obvious way to control product purity over changes in the vapor feed concentration.

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# VIII. APPENDIX

A. Summary of Data and Calculated Values.

# E<sub>b</sub> Series 2

# PLATE NUMBER





Time Constant, Series 2



# E p<sup>Series</sup> 3

# PLATE NUMBER



 $\mathcal{R}$ 

# E<sub>D</sub> Series 3 (continued)

PLATE NUMBER



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# $E_D$ , Series 4<br>(continued)

# PLATE NUMBER



# $E_D$ , Series 4 (continued)

# PLATE NUMBER



# Lag Time, Series 7

# Plate

Step<br>(0.07 to -)



 $\mathbb{E}_{\mathbb{E}_{\mathcal{B}}}$ 

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# VIII. APPENDIX

# B., Location of Original Data

The Output Table graphs for each run in all of the Series are included in the third copy of this thesis. This copy in in the possession of Professor E.R.Gilliland Department of Chemical Engineering, Massachusetts Institute of Technology.

Numerical data and calculations of the original work may be found in the research notebook "Morris-1960" which 1s located at the Department of Chemical Engineering, Massachusetts Institute of Technology.

# VIII. APPENDIX

# C. Nomenclature





 $\sim$ r

# Subscripts



liquid feed

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#### V. APPENDIX

#### D. Literature Citations

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